

Control Systems I

Recitation 04

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Last Week

System Classification

Linear → Superposition possible $\Sigma(\alpha \cdot u_1 + \beta \cdot u_2) = \alpha \cdot \Sigma(u_1) + \beta \cdot \Sigma(u_2)$ $y(t) = \alpha \cdot u(t), \quad y(t) = \frac{d}{dt}u(t), \quad y(t) = \int_0^t u(\tau)d\tau$	Non-Linear $\Sigma(\alpha \cdot u_1 + \beta \cdot u_2) \neq \alpha \cdot \Sigma(u_1) + \beta \cdot \Sigma(u_2)$ $y(t) = \alpha \cdot u(t) + \beta, \quad y(t) = \sin(u(t))$
Causal = <i>not</i> dependant on <i>future</i> values $y(t) = u(t - \tau) \quad \forall \tau \geq 0, \quad y(t) = \int_{-\infty}^0 u(\tau)d\tau$	Non-Causal = dependant on <i>future</i> values $y(t) = u(t + 5), \quad \int_{-\infty}^{t+1} u(\tau)d\tau$
Static = dependant on current values only $y(t) = 3 \cdot u(t), \quad y(t) = \sqrt{u(t)}$	Dynamic = also dependant on past values. $y(t) = \int_0^t u(\tau)d\tau, \quad y(t) = u(t - \tau) \quad \forall \tau \neq 0$
Time invariant $y(t) = \frac{d}{dt}u(t), \quad y(t) = 3 \cdot u(t)$	Time varying = parameters are time dependant $y(t) = \sin(t) \cdot u(t), \quad y(t) = u(t) + t$
State $x(t)$ Vector $x(t) \in \mathbb{R}^n$ of values at t that fully describe the system. Past and future!	Dimension Minimal number of variables, n , to fully describe the system.

Last Week

LTI-Systems

- LTI state-space model: notice A, B, C and D are constant

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

$$A = \left. \frac{\partial f(x, u)}{\partial x} \right|_{(x_e, u_e)} = \left[\begin{array}{ccc} \frac{\partial f_1}{\partial x_1} & \cdots & \frac{\partial f_1}{\partial x_n} \\ \vdots & \ddots & \vdots \\ \frac{\partial f_n}{\partial x_1} & \cdots & \frac{\partial f_n}{\partial x_n} \end{array} \right]_{(x_e, u_e)} \in \mathbb{R}^{n \times n}$$

$$B = \left. \frac{\partial f(x, u)}{\partial u} \right|_{(x_e, u_e)} = \left[\begin{array}{ccc} \frac{\partial f_1}{\partial u_1} & \cdots & \frac{\partial f_1}{\partial u_m} \\ \vdots & \ddots & \vdots \\ \frac{\partial f_n}{\partial u_1} & \cdots & \frac{\partial f_n}{\partial u_m} \end{array} \right]_{(x_e, u_e)} \in \mathbb{R}^{n \times m}$$

$$C = \left. \frac{\partial g(x, u)}{\partial x} \right|_{(x_e, u_e)} = \left[\begin{array}{ccc} \frac{\partial g_1}{\partial x_1} & \cdots & \frac{\partial g_1}{\partial x_n} \\ \vdots & \ddots & \vdots \\ \frac{\partial g_p}{\partial x_1} & \cdots & \frac{\partial g_p}{\partial x_n} \end{array} \right]_{(x_e, u_e)} \in \mathbb{R}^{p \times n}$$

$$D = \left. \frac{\partial g(x, u)}{\partial u} \right|_{(x_e, u_e)} = \left[\begin{array}{ccc} \frac{\partial g_1}{\partial u_1} & \cdots & \frac{\partial g_1}{\partial u_m} \\ \vdots & \ddots & \vdots \\ \frac{\partial g_p}{\partial u_1} & \cdots & \frac{\partial g_p}{\partial u_m} \end{array} \right]_{(x_e, u_e)} \in \mathbb{R}^{p \times m}$$

- Remember this is always a deviation from the equilibrium point:

Outline

- Time Response
 - What? Why? How?
 - Example
- Similarity Transform
 - What? Why? How?
 - Example
 - Connection to Stability
- Stability
 - Definition
 - How?
 - Linear and Non-Linear Systems

Time Response

What? Why?

- What?
 - Find a solution to the LTI system

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

- $y(t) = S(u(t), x_0, A, B, C, D)$
- Why?
 - Test different Inputs (as last week)
 - Understand/Derive the notion of Stability
 - Groundwork for Frequency Domain calculations

Time Response

How?

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

- We can use superposition(linearity) and look at two cases:
 - 1. No input and any initial condition: $u_{IC}(t) = 0, x_{0,IC} = x_0$
 - 2. Any input and zero initial condition: $u_F(t) = u(t), x_{0,F} = 0$
 - They satisfy $x_0 = x_{0,IC} + x_{0,F}$ and $u(t) = u_{IC}(t) + u_F(t)$ thus we have $y(t) = y_{IC}(t) + y_F(t)$
- Looking at the zero input we have (Initial Condition Response):
 - $\frac{d}{dt}x_{IC}(t) = A \cdot x_{IC}(t)$
 - Easy to solve: $x_{IC}(t) = e^{At}x_0$
- Looking at the zero initial condition we have (Forced Response):
 - $\frac{d}{dt}x_F(t) = A \cdot x_F(t) + B \cdot u(t)$
 - Not so easy put can be shown that: $x_F(t) = \int_0^t e^{A(t-\tau)}Bu(t)d\tau$

Time Response

How?

- Using the definition of $y(t)$ and the previous solutions we get:
 - $y(t) = C \cdot e^{At} \cdot x_0 + C \cdot \int_0^t e^{A(t-\tau)} B u(\tau) d\tau + D \cdot u(t)$
 - Initial Condition Response:
 - $y_{IC}(t) = C \cdot e^{At} \cdot x_0$
 - Describes how the system behaves naturally
 - Force Response:
 - $y_F(t) = C \cdot \int_0^t e^{A(t-\tau)} B u(\tau) d\tau$
 - How the system reacts to the input
 - Feedthrough:
 - $y_{FF}(t) = D \cdot u(t)$
 - Direct effect of the input (usually 0)

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

Time Response

How?

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

- We have:

- $y(t) = C \cdot e^{At} \cdot x_0 + C \cdot \int_0^t e^{A(t-\tau)} B u(\tau) d\tau + D \cdot u(t)$

- Matrix Exponential:

- $e^{At} = \sum_{k=0}^{\infty} \frac{(At)^k}{k!} = I + At + \frac{1}{2}(At)^2 + \dots + \frac{1}{k!}(At)^k + \dots$

- $e^{At} = \begin{bmatrix} e^{A_{11}t} & 0 \\ 0 & e^{A_{22}t} \end{bmatrix}$, if A is diagonal

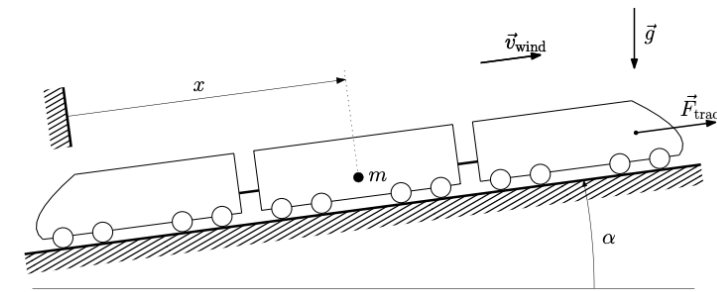
- Jordan Form:

- $A = \begin{bmatrix} \lambda & 1 \\ 0 & \lambda \end{bmatrix} \rightarrow e^{At} = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} e^{\lambda t}$

- $A = \begin{bmatrix} \lambda & 1 & 0 \\ 0 & \lambda & 1 \\ 0 & 0 & \lambda \end{bmatrix} \rightarrow e^{At} = \begin{bmatrix} 1 & t & \frac{1}{2!}t^2 \\ 0 & 1 & t \\ 0 & 0 & 1 \end{bmatrix} e^{\lambda t}$

- See Similarity Transform to get these matrices
 - Keep in mind in practice we never calculate this by hand (useful for theory)

Time Response



Example - Train Driving up a Hill ($v_{\text{wind}} = v_{\text{wind}}(t)$, $\alpha = \text{const}$)

- We got the LTI system with $u(t) = u_0 = \text{const.}$, $x_0 = [0, 0]^T$:
 - $\frac{d}{dt}x(t) = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{1}{m}\rho c_w A v_{\text{wind}} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix} u(t) = \begin{bmatrix} 0 & 1 \\ 0 & -L \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ b \end{bmatrix} u(t)$
 - $y(t) = [1 \quad 0] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + 0 u(t)$
 - Using: $y(t) = C \cdot e^{At} \cdot x_0 + C \cdot \int_0^t e^{A(t-\tau)} B u(\tau) d\tau + D \cdot u(t)$
- We have: $e^{At} = \begin{bmatrix} 1 & -\frac{e^{-Lt}-1}{L} \\ 0 & e^{-Lt} \end{bmatrix}$
 - $y(t) = [1 \quad 0] \begin{bmatrix} 1 & -\frac{e^{-Lt}-1}{L} \\ 0 & e^{-Lt} \end{bmatrix} \begin{bmatrix} 0 \\ 0 \end{bmatrix} + [1 \quad 0] \cdot \int_0^t \begin{bmatrix} 1 & -\frac{e^{-L(t-\tau)}-1}{L} \\ 0 & e^{-L(t-\tau)} \end{bmatrix} \begin{bmatrix} 0 \\ b \end{bmatrix} u_0 d\tau$
 - $y(t) = 0 + \int_0^t -\frac{e^{-L(t-\tau)}-1}{L} b u_0 d\tau = -\frac{b u_0}{L} \int_0^t e^{-L(t-\tau)} - 1 d\tau = \frac{b}{L^2} (e^{-Lt} - 1 + Lt)$
 - $y(t) = \frac{b}{L^2} (e^{-Lt} - 1 + Lt)$
 - If we would have chosen $C = [0, 1]$ (only velocity as output)
 - $y(t) = \frac{b}{L} (1 - e^{-Lt})$

Similarity Transform

What? Why?

- What?
 - Rewrite the state of our system $\tilde{x}(t) = Kx(t)$

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

- Not so relevant for exam but good to know/useful trick
- Why?
 - Sometimes easier system to work with
 - Get minimal realisation
 - **Diagonalize the system**
 - Matrix Exponential

Similarity Transform

$$\begin{aligned}\frac{d}{dt}x(t) &= A \cdot x(t) + B \cdot u(t) \\ y(t) &= C \cdot x(t) + D \cdot u(t)\end{aligned}$$

How?

- Define an invertible matrix T :
 - $x(t) = T \cdot \tilde{x}(t) \rightarrow \tilde{x}(t) = T^{-1}x(t)$
 - Our system then becomes:
 - $T \frac{d}{dt} \tilde{x}(t) = A \cdot T \cdot \tilde{x}(t) + B \cdot u(t) \Rightarrow \frac{d}{dt} \tilde{x}(t) = T^{-1} \cdot A \cdot T \cdot \tilde{x}(t) + T^{-1} \cdot B \cdot u(t)$
 $y(t) = C \cdot T \cdot \tilde{x}(t) + D u(t) \qquad y(t) = C \cdot T \cdot \tilde{x}(t) + D u(t)$
- This results in the system:
 - $\frac{d}{dt} \tilde{x}(t) = \tilde{A} \cdot \tilde{x}(t) + \tilde{B} \cdot u(t)$, with $\tilde{A} = T^{-1}AT$, $\tilde{B} = T^{-1}B$, $\tilde{C} = CT$
 $y(t) = \tilde{C} \cdot \tilde{x}(t) + \tilde{D} u(t)$
- It can be shown that these systems are the same

Similarity Transform

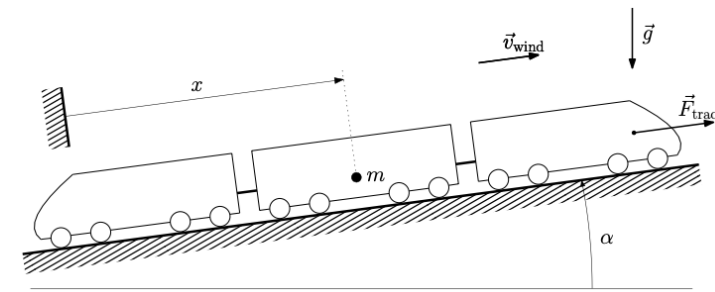
How?

- Most often we use Eigenvalue Transform (Eigendecomposition):
 - Recall Eigenvalues: $Av_i = \lambda_i v_i$
 - Put that into a matrix: V is the matrix of Eigenvectors
 - $AV = A[v_1, v_2, \dots, v_n] = [\lambda_1 v_1, \lambda_2 v_2, \dots, \lambda_n v_n] = V\Lambda$, $\Lambda = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_n)$
 - $V^{-1}AV = \Lambda$
 - This looks very familiar:
 - $\frac{d}{dt}\tilde{x}(t) = \tilde{A} \cdot \tilde{x}(t) + \tilde{B} \cdot u(t)$, with $\tilde{A} = T^{-1}AT$, $\tilde{B} = T^{-1}B$, $\tilde{C} = CT$
 $y(t) = \tilde{C} \cdot \tilde{x}(t) + \tilde{D} u(t)$
- We have $T = V$ and $\tilde{A} = \Lambda$, $\tilde{B} = V^{-1}B$, $\tilde{C} = CV$, $x = V\tilde{x} \rightarrow \tilde{x} = V^{-1}x$
 - with $V = [v_1, v_2, \dots, v_n]$ and $\Lambda = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_n)$
- For the output we have:
 - $y(t) = \tilde{C} \cdot e^{\tilde{A}t} \cdot \tilde{x}_0 + \tilde{C} \cdot \int_0^t e^{\tilde{A}(t-\tau)} \tilde{B}u(\tau)d\tau + \tilde{D} \cdot u(t)$
 - $y(t) = C \cdot V \cdot e^{\Lambda t} \cdot V^{-1}x_0 + C \cdot V \cdot \int_0^t e^{\Lambda(t-\tau)} V^{-1}Bu(\tau)d\tau + D \cdot u(t)$

Similarity Transform

Example - Train Driving up a Hill ($v_{\text{wind}} = v_{\text{wind}}(t)$, $\alpha = \text{const}$)

- Recall the system with $u(t) = u_0 = \text{const.}$, $x_0 = [0, 0]^T$:
 - $\frac{d}{dt}x(t) = \begin{bmatrix} 0 & 1 \\ 0 & -L \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ b \end{bmatrix} u(t)$
 - $y(t) = [1 \quad 0] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + 0 u(t)$
- Get eigenvalues and Eigenvectors of A:
 - $V = \begin{bmatrix} 1 & -\frac{1}{L} \\ 0 & 1 \end{bmatrix}, \Lambda = \begin{bmatrix} 0 & 0 \\ 0 & -L \end{bmatrix}, V^{-1} = \begin{bmatrix} 1 & \frac{1}{L} \\ 0 & 1 \end{bmatrix}$
 - $y(t) = C \cdot V \cdot e^{\Lambda t} \cdot V^{-1} x_0 + C \cdot V \cdot \int_0^t e^{\Lambda(t-\tau)} V^{-1} B u(t) d\tau + D \cdot u(t)$
 - $y(t) = 0 + [1 \quad 0] \begin{bmatrix} 1 & -\frac{1}{L} \\ 0 & 1 \end{bmatrix} \int_0^t \begin{bmatrix} 1 & 0 \\ 0 & e^{-L(t-\tau)} \end{bmatrix} d\tau \begin{bmatrix} 1 & \frac{1}{L} \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ b \end{bmatrix} u_0$
 - $y(t) = 0 + [1 \quad -\frac{1}{L}] \begin{bmatrix} t & 0 \\ 0 & \frac{1-e^{-Lt}}{L} \end{bmatrix} \begin{bmatrix} \frac{1}{L} b \\ b \end{bmatrix} u_0 = \frac{b}{L^2} (e^{-Lt} - 1 + Lt)$
 - We see same as before!!



Similarity Transform

Connection to Stability

- We had the Eigenvalue Transform (Eigendecomposition):
 - $\tilde{A} = \Lambda, \tilde{B} = V^{-1}B, \tilde{C} = CV, x = V\tilde{x} \rightarrow \tilde{x} = V^{-1}x$
 - with $V = [v_1, v_2, \dots, v_n]$ and $\Lambda = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_n)$
 - $y(t) = C \cdot V \cdot e^{\Lambda t} \cdot V^{-1}x_0 + C \cdot V \cdot \int_0^t e^{\Lambda(t-\tau)} V^{-1}Bu(\tau)d\tau + D \cdot u(t)$
- For a system with no input (natural response):
 - If only real Eigenvalues:
 - $e^{\Lambda t} = \begin{bmatrix} e^{\lambda_1 t} & 0 \\ 0 & e^{\lambda_2 t} \end{bmatrix}$
 - some sort of exponential decay/growth
 - In the case of imaginary Eigenvalues/Eigenvectors: $\lambda = \sigma + j\omega$
 - $e^{\Lambda t} = \begin{bmatrix} e^{(\sigma+j\omega)t} & 0 \\ 0 & e^{(\sigma+j\omega)t} \end{bmatrix} = \begin{bmatrix} e^{\sigma t} \sin(\omega t + \phi) & 0 \\ 0 & e^{\sigma t} \sin(-\omega t + \phi) \end{bmatrix}$
 - some sort of exponential decay/growth with oscillation
 - This gives us a hint for Stability
 - Keep in mind this may be some “unreal” state but the qualitative behavior is the same

Stability

Definition

- Lyapunov:
 - Lyapunov Stable:
 - For no input and a bounded initial condition the system remains bounded
 - $\|x_0\| < \epsilon$, and $u = 0 \rightarrow \|x(t)\| < \delta, \forall t \geq 0$
 - Lyapunov Asymptotically Stable:
 - For a bounded initial condition and no input the system converges to 0
 - $\|x_0\| < \epsilon$, and $u = 0 \rightarrow \lim_{t \rightarrow \infty} \|x(t)\| = 0$
 - Lyapunov Unstable: If not stable or asymptotically stable
- BIBO:
 - BIBO (Bounded Input, Bounded Output):
 - For a zero initial condition and bounded input the systems output remains bounded
 - $\|u(t)\| < \epsilon \quad \forall t \geq 0$, and $x_0 = 0 \rightarrow \|y(t)\| < \delta, \quad \forall t \geq 0$

Stability

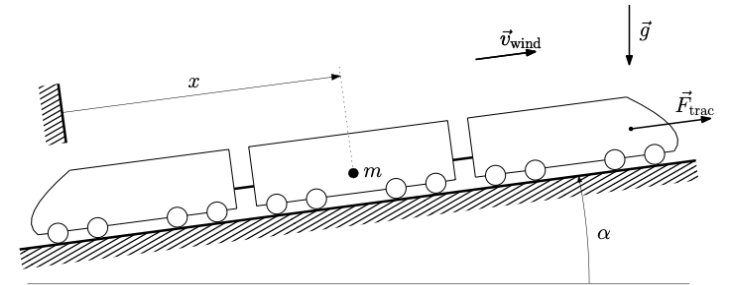
How?

- We saw that the Eigenvalues tell us a lot about the system!!!
- 1. Calculate the Eigenvalues of the A matrix. $\det(A - I\lambda) = 0$
- 2. Lyapunov:
 - Lyapunov Stable:
 - A system is Lyapunov stable if $\text{Re}(\lambda_i) \leq 0, \forall \lambda_i$
 - All Eigenvalues of A have real-part less or equal to zero
 - Lyapunov Asymptotically Stable
 - A system is Lyapunov asymptotically stable if $\text{Re}(\lambda_i) < 0, \forall \lambda_i$
 - All Eigenvalues of A have real-part less than zero
 - Lyapunov Unstable
 - A system is Lyapunov unstable stable if $\text{Re}(\lambda_i) > 0$, for any λ_i
- 3. BIBO:
 - BIBO Stable:
 - A minimal LTI system is BIBO stable if it is Lyapunov Asymptotically Stable

Stability

Linear and Non-Linear systems

- This notion of stability only holds for the linear system!!!
- For the non-linear system we have:
 - If Lyapunov unstable -> Non-linear system is unstable
 - If Lyapunov asymptotically stable -> Non-linear system is stable
 - If Lyapunov system is stable -> nothing can be said (non-linear systems theory is needed)
- For the train example:
 - $A = \begin{bmatrix} 0 & 1 \\ 0 & -L \end{bmatrix}$
 - $\Lambda = \begin{bmatrix} 0 & 0 \\ 0 & -L \end{bmatrix} \rightarrow \lambda_1 = 0, \lambda_2 = -L$
 - System is Lyapunov stable
 - Non-linear system:
 - We can't know, from the linear system itself.
 - Depends on α, v_{wind} etc.



Stability

Example – Old Exam question

- We have a matrix A:

- $A = \begin{bmatrix} 1 & -2 \\ \alpha & -3 \end{bmatrix}$

- For what α is the system asymptotically stable?

$$\det(\lambda I - A) = \det \left(\begin{bmatrix} \lambda & 0 \\ 0 & \lambda \end{bmatrix} - \begin{bmatrix} 1 & -2 \\ \alpha & -3 \end{bmatrix} \right)$$

$$= \det \left(\begin{bmatrix} \lambda - 1 & 2 \\ -\alpha & \lambda + 3 \end{bmatrix} \right)$$

$$= (\lambda - 1)(\lambda + 3) + 2\alpha$$

$$= \lambda^2 + 2\lambda + (2\alpha - 3) \stackrel{!}{=} 0$$

- $\lambda_{1,2} = \frac{(-2 \pm \sqrt{4 - 4(2\alpha - 3)})}{2} = \frac{(-2 \pm \sqrt{4 - 8\alpha + 12})}{2} = \frac{(-2 \pm \sqrt{16 - 8\alpha})}{2}$

- $\operatorname{Re}(\lambda_1) < 0$ for any α

- $\operatorname{Re}(\lambda_2) < 0$:

- $-2 + \sqrt{16 - 8\alpha} < 0 \rightarrow 16 - 8\alpha < 4 \rightarrow 12 < 8\alpha$

- $\operatorname{Re}(\lambda_2) < 0$: for $\alpha > 1.5$

