

## Signal and System Theory II

This sheet is provided to you for ease of reference only.  
*Do not write your solutions here.*

### Exercise 1

1	2	3	4	5	Exercise
6	6	1	6	6	25 Points

A loudspeaker and its driving circuitry are shown schematically in Figure 1. The horizontal displacement of the loudspeaker cone with equivalent mass of  $m$  is denoted by  $z(t)$ .  $R$  denotes the resistance of the loudspeaker coil, while  $L$  the coil's inductance.

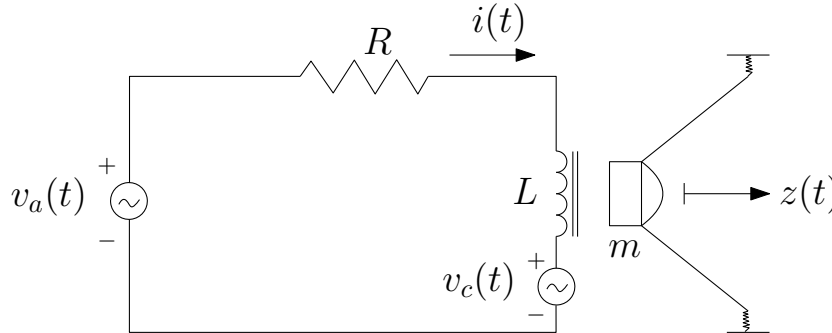


Figure 1: Schematic of a simplified electromechanical model of a loudspeaker.

The loudspeaker cone is driven by the force

$$F(t) = \ell i(t), \quad (1)$$

where  $i(t)$  is the current through the coil and  $\ell$  a coil parameter. The coil voltage  $v_c(t)$  is related to the motion of the loudspeaker cone and is given by

$$v_c(t) = \ell \dot{z}(t). \quad (2)$$

The air resistance to the cone movement is proportional to the cone velocity with coefficient  $d$ . Assume that all parameters are positive  $m, R, L, \ell, d > 0$  and that the vertical displacement of the cone and all other forces acting on it are negligible.

1. Choosing as state  $x(t) = [z(t) \quad \dot{z}(t) \quad i(t)]^T$ , the applied voltage  $v_a(t)$  as input and the cone displacement  $z(t)$  as output show that the state-space model for the system can be described by the following system matrices.

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & -\frac{d}{m} & \frac{\ell}{m} \\ 0 & -\frac{\ell}{L} & \frac{R}{L} \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}, \quad C = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}, \quad D = \begin{bmatrix} 0 \end{bmatrix}.$$

Matrix B is missing a 1/L, please keep that in mind when solving the exercise.

2. Is the system observable? Is it controllable?
3. Calculate the equilibrium of the system under a constant zero input voltage  $v_a(t) = 0$ .
4. State the conditions on the parameter values under which the system is stable and asymptotically stable.
5. Consider now a state feedback controller of the form

$$u(t) = K x(t) = \begin{bmatrix} k_1 & \frac{\ell}{L} & k_3 \end{bmatrix} x(t). \quad (3)$$

Derive conditions on  $k_1$  and  $k_3$  such that the closed-loop system  $\dot{x} = (A + BK)x(t)$  is asymptotically stable.

**Hint:** The third-order polynomial  $P(s) = s^3 + a_2 s^2 + a_1 s + a_0$  has roots with negative real part if and only if its coefficients are positive ( $a_0, a_1, a_2 > 0$ ) and  $a_2 a_1 > a_0$ .

## Exercise 2

1	2	3	4	5	Exercise
4	4	7	5	5	25 Points

Consider the following linear time invariant system with parameters  $a, b, c \in \mathbb{R}$ :

$$\begin{aligned}\dot{x}(t) &= \overbrace{\begin{bmatrix} -1 & 4 \\ a & -2 \end{bmatrix}}^A x(t) + \overbrace{\begin{bmatrix} 0 \\ b \end{bmatrix}}^B u(t) \\ y(t) &= \underbrace{\begin{bmatrix} c & 1 \end{bmatrix}}_C x(t).\end{aligned}$$

1. For what values of  $(a, b, c)$  does the equation  $A^T P + P A = -I$  have a unique symmetric positive definite solution  $P$ ? What does this imply for the zero-input-response of the system?
2. For what values of  $(a, b, c)$  is the system controllable?

For the remainder of the exercise, assume  $(a, b, c) = (3, 1, 1)$ .

3. Consider a feedback controller of the form

$$u(t) = Kx(t) = \begin{bmatrix} 0 & k \end{bmatrix} x(t).$$

Assume you want the closed-loop eigenvalues,  $\lambda_1$  and  $\lambda_2$ , of  $\dot{x}(t) = (A + BK)x(t)$  to satisfy  $\lambda_2 = \alpha\lambda_1$ , for  $\lambda_1 < 0$  and  $\alpha > 0$ . For which values of  $\alpha > 0$  can we always find a  $k \in \mathbb{R}$  so that this condition is satisfied?

**Hint:** Write down the characteristic polynomial of  $(A + BK)$  and compare it with a polynomial whose roots are  $\lambda_1$  and  $\alpha\lambda_1$ . Remember that  $k$  is a real number and a quadratic with real coefficients has real roots if and only if its discriminant is non-negative. You may assume that  $0.5\sqrt{2496} \approx 24.98$ .

4. Take now  $\alpha = 0.02$  in Part 3. Find the  $k$  that leads to the fastest response of the closed loop system  $\dot{x}(t) = (A + BK)x(t)$ .
5. You would like to implement the controller from Part 4 when you are only able to measure the output  $y(t)$  instead of the whole state  $x(t)$ . Is it possible to design an observer to infer the states from the output? You may answer without deriving the equations for the observer.

### Exercise 3

1(a)	1(b)	2	Exercise
5	10	10	25 Points

1. Consider the system whose block diagram is shown in Figure 2. Assume  $H(s) = \frac{1}{s-2}$ ,  $F(s) = \frac{as+5}{s+2}$ , and  $a \in \mathbb{R}$ .

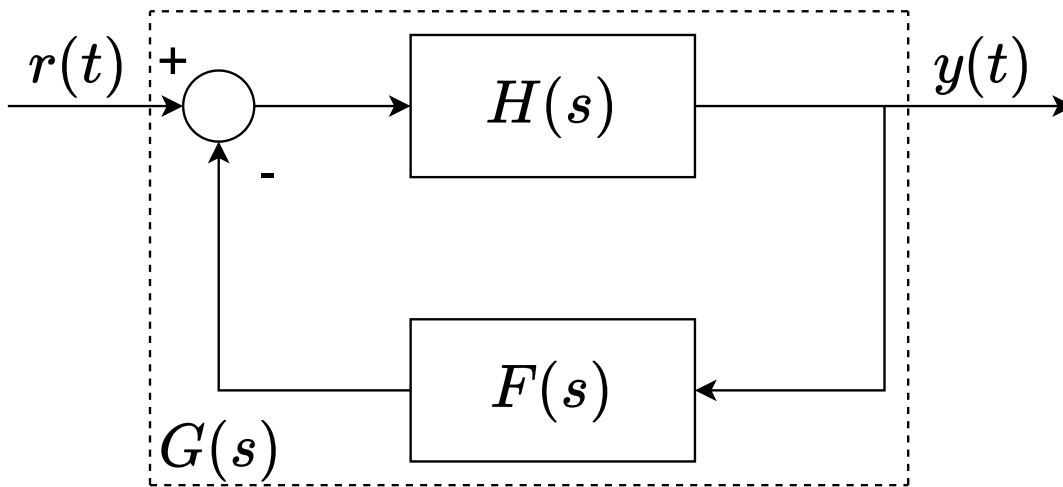


Figure 2: Block diagram.

- (a) Verify that the transfer function  $G(s)$  from the input  $r(t)$  to the output  $y(t)$  is given by

$$G(s) = \frac{s+2}{s^2 + as + 1}$$

- (b) Consider  $a \in \{-2, 0, 0.5, 2\}$ . Match the step responses of  $G(s)$  in Figure 3 with the corresponding value of the parameter  $a$ . Justify your answer.

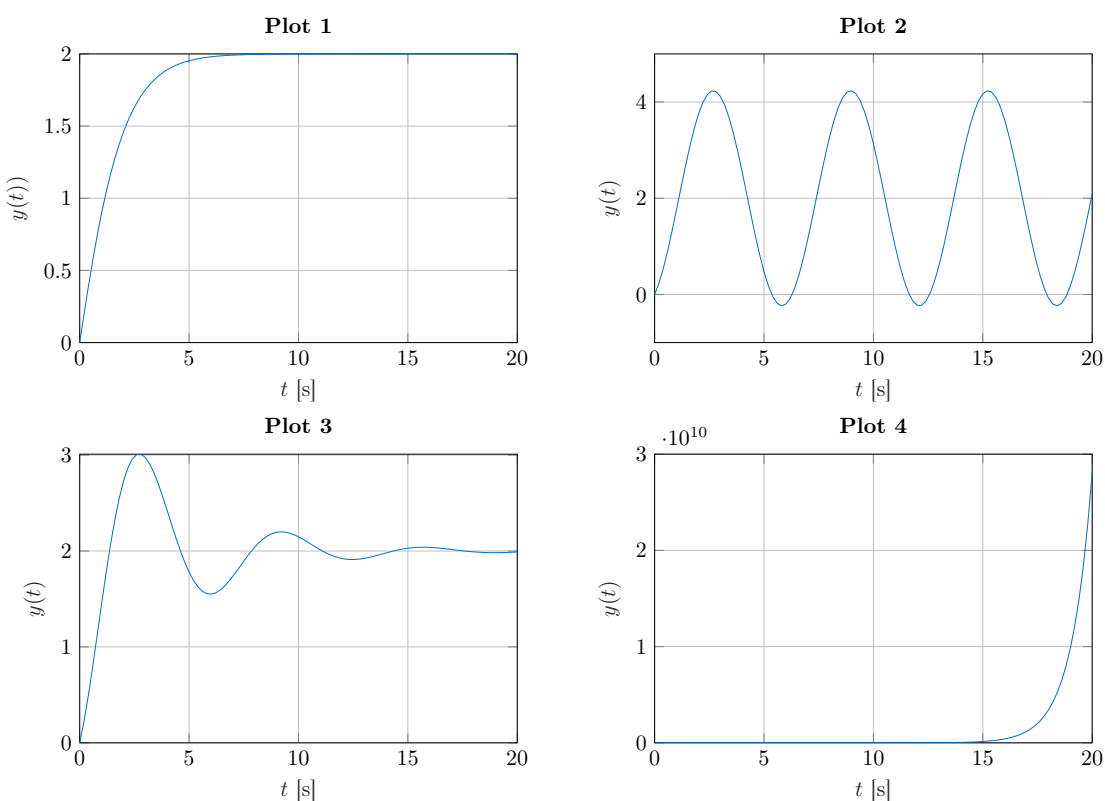


Figure 3: Step responses of the system  $G(s)$  for different values of the parameter  $a$ .

- The system  $G(s)$  in Part 1 with  $a = -2$  is to be controlled via a proportional gain  $K \in \mathbb{R}$  as in Figure 4. Given the Nyquist plot in Figure 5, for which values of  $K$  will the closed loop system be stable?

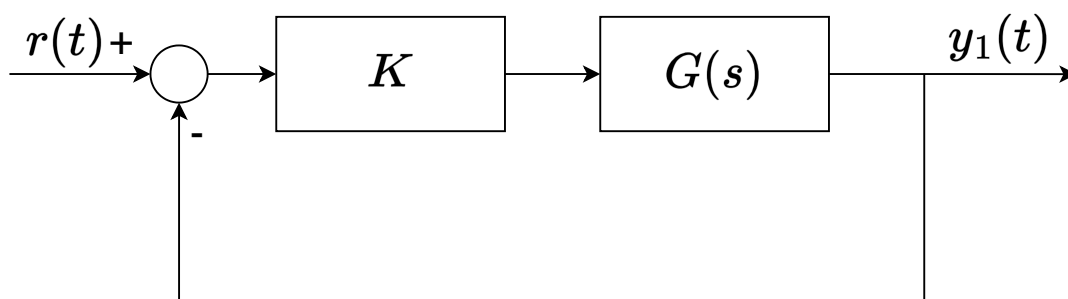


Figure 4: Block diagram.

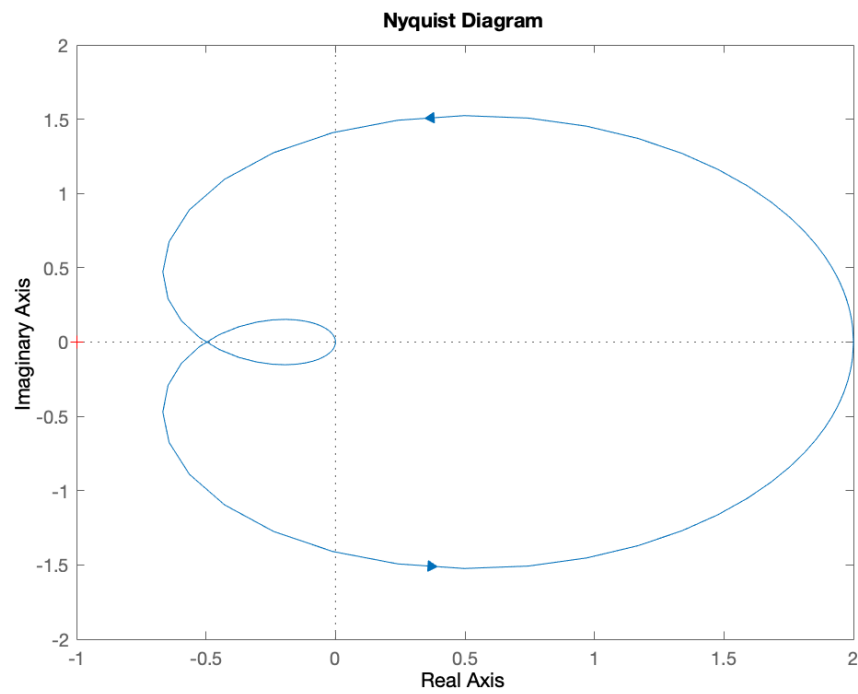


Figure 5: Nyquist plot of the system  $G(s)$  for  $a = -2$ .

## Exercise 4

1	2	3	4	5	Exercise
6	1	5	8	5	25 Points

Consider the system given by the following state-space equations

$$\begin{aligned}\dot{x}_1(t) &= x_1(t)x_2(t), \\ \dot{x}_2(t) &= -kx_2(t) - x_1^2(t) + a\end{aligned}$$

with state variables  $x_1(t), x_2(t) \in \mathbb{R}$  and constants  $a, k \in \mathbb{R}$ .

1. Derive all equilibrium points of the system. How many equilibrium points does the system have for  $a > 0$  and  $a < 0$ ?
2. Verify that when  $a = 0$ ,  $\hat{x} = (0, 0)$  is the unique equilibrium of the system.
3. For  $a = 0$  what can you say about the stability of the equilibrium  $\hat{x} = (0, 0)$  for different values of  $k \in \mathbb{R}$  using linearisation?
4. Can you use the Lyapunov function  $V(x) = \frac{1}{2}(x_1^2 + x_2^2)$  to show stability of the equilibrium  $\hat{x} = (0, 0)$  for  $a = 0$  and  $k > 0$ ? If yes, prove stability, if not justify your answer! Can you use the same Lyapunov function to prove local asymptotic stability for the same values of  $a$  and  $k$ ?

**Hint:** You may take  $S = \mathbb{R}^2$  as the open set used in the theorem.

5. For  $K \geq 0$  consider the set  $S_K = \{x(t) \in \mathbb{R}^2 \mid V(x) \leq K\}$ . Using the Lyapunov function  $V(x)$  from Part 4, argue that the set  $S_K$  is invariant when  $a = 0$  and  $k > 0$  for any  $K \geq 0$ . Hence, using LaSalle's theorem, show that the equilibrium  $\hat{x} = (0, 0)$  is globally asymptotically stable.

**Hint:** You may assume that the set  $S_K$  is compact for all  $K \geq 0$ .